# A MODERN SELF QUANTIZING TECHNIQUE ARCHITECTURES USING DISCRETE WAVELET TRANSFORM FOR IMAGE PROCESSING APPLICATIONS

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#### **ABSTRACT**

This paper exists Precision-Aware approaches and related hardware implementations for performing the operation of DWT. It includes implementation of BP architecture and DS design methodologies. These methodologies will give the optimal utilization of hardware resources in the computation of Discrete Wavelet Transform (DWT). Moreover, quantization of DWT coefficients to a precise target step size. It is performed as an inbuilt part of the DWT computation, so eliminating the need to have a divide downstream quantization step in applications like JPEG 2000. It concludes that DS design requires extensively less hardware resources with increasing accuracy and DWT level when BS design exhibit natural speed is advances.

# Keywords — Digit Serial, Bit Parallel, Lifting Based DWT, Quantization, DWT

# I. INTRODUCTION

Discrete wavelet transform (DWT) splits an image into multiple and small bands (sub bands) of low and high frequency components. To compress an image, we need to encode these sub bands. This encoding technique and main DWT technique gives us the total information of an image with less bits. Image compression is applicable in the fields of entertainment, medical, defense, commercial, industrial domains and etc. DWT plays an important role in image. DWT and Inverse DWT for Transformations are using in other image techniques like image enhancement, image restoration and image filtering. DWT-IDWT is one of the well-known conversion techniques that are extensively used in signal processing and communication applications. Discrete and inverse discrete wavelet techniques are transforms signal into multiple promise sub bands. DWT is computationally very exhaustive and requires power due to more number of arithmetical operations.

Latency and throughput are other foremost margins of DWT and it has compound levels of hierarchy. DWT was conventionally implemented by convolution technique. The architecture complexity is depends upon the serial or parallel digit representation of input data. As mentioned as an implementation requires large amount of storage and large number of computations. It is for either low-power or high-speed applications. Newly, a lifting-based scheme has been proposed for the DWT, which requires less computation. The main aspects of the lifting based DWT scheme is to disintegrate high pass filters into a sequence of upper triangular matrices, low

pass filters into a sequences of lower triangular matrices and translate the filter implementation into stripy matrix.

The JPEG 2000 standard has good advantages over the original block DCT-based JPEG standard such as significant coding efficiency and flexibility. The important element of JPEG 2000 is the discrete wavelet transform (DWT), which recursively decomposes an input image into sub bands. Those sub bands has different spatial frequency and point of reference. The most frequently used DWT filters in JPEG 2000 are the biorthogonal I) lossless 5/3 integer and II) lossy 9/7 floating-point filter banks. In this document, we keep focus on the DWT using 9/7 filter, which provides very good compression quality. But it is particularly challenging to implement with high efficiency because of the reason is to the irrational personality of the filter coefficients.

#### II. LITERATURE SURVEY

The committee of JPGE has begun the investigation for finding another type of image compression technique which will be useful for current and future applications. The committee has found one technique named JPGE 2000, is a compressive technique. It not only has higher compression efficiency compared to other systems like baseline JPGE system but also provides new rich set of representation. In this technique, memory efficient block DCT of JPGE has replaced by the full frame DWT (Discrete Wavelet Transform) and has low-complexity. The DWT improves compression efficiency because of the reason is good energy compaction and also provides image representation of multi-resolution.

There are so many themes on the different types of hardware implementations of the DWT algorithm and the novel DWT algorithm. Those articles have paid very less immersion to the meticulousness computation of the DWT. We can consider it as a design goal. Some articles are comparatively comprises few treatments for this problem. The work in "Quantization effect on VLSI implementations for the 9/7 DWT filters" presents a novel architecture for 1-D and 2-D DWT by using lifting schemes. To achieve the goal of low and high - frequency component of original data being available alternately, this has designed. The effects of quantizing the lifting coefficients of the 9/7 DWT has achieved in the previous article. The possessions on the peak signal-to-noise ratio (PSNR) and hardware area/speed are evaluated from the number of canonical signed digit (SD) terms for the coefficients are varied. After this another article has conducted a similar analysis but it is with the fixed – point data path fixed to 12 bits of integer and the same bits of fractional accuracy. It provides adequate dynamic range, which is useful to compute a six-level DWT over 50 dB PSNR.

In difference to the preceding work, which has been mainly directed to filter coefficients, in this paper, we lecture to synchronized optimization of not only the coefficient precision but also the internal data paths, which have used in their computation and for that purpose we present a solution that is fully generalized with stare to accuracy, allowing design of a DWT to any desired precision. The optimization technique can be able to use for minimization of the operand bit width in a bit parallel (BP) architecture and also to use for minimization of iterations in digit- serial (DS) architecture. This facilitates implementations with a considerable improvement in hardware resources and/or execution times while also make a sure that overflows are avoided and requirements of precision should be met.

This presentation includes sections as follows: Top level architecture of the DWT design in Section III, lifting-based DWT and JPEG2000 quantization in discussed in Section IV, proposed design has explained in the section V, DS-DWT has explained in Section VI, and finally Section VII will give you the result of our design.

#### III. OP LEVEL ARCHITECTURE

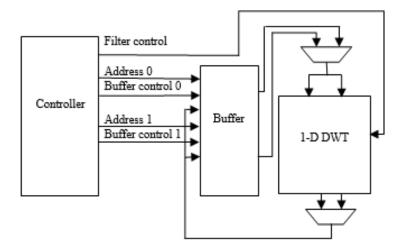


Fig 1: Generic High-Level Architecture of the DWT Design

The figure represents the generic high level architecture of the DWT design. In figure, the 1-D DWT module performs the actual operation of DWT, which can be implemented by bit parallel (BP) architecture. Here the buffer is a dual-port. It holds two data frames, which can be used to store the original raw data, intermediate data, and /or the final transformed data. Another important module: the controller is used to control the whole operation of design. It has done by creating control signals for the buffer and also for the filter.

# IV. LIFTING BASED DWT

#### 4.1 Lifting

The lifting scheme based DWT has been integrated in the forthcoming JPEG2000 standard because it diminishes the arithmetic complexity of the conventional as compared with the convolution based DWT.

Fig.2 gives you an idea about the procedure for performing a two level DWT on an image. The 1-D DWT carry out the actual wavelet transform by first performing the row wise operation on the image to produce low frequency L1 and high- frequency H1 components. After performing 1-D DWT again on the columns of L1 and H1, the first level of decomposition has to complete on those components, then we got LL1, HL1, LH1, and HH1. This process can be repeated on LL1 to produce the LL2, HL2, LH2, and HH2 sub bands.

Actually the 9/7 DWT was implemented depending on convolution-based methods. Generally low pass and high pass filters are used in convolution-based methods. Afterwards Daubenchies and Sweldens proved that DWT can be decomposed into a finite sequence of lifting steps. It provides several advantages like lower computation and memory requirements and also easier boundary management explained in "A survey on lifting-based discrete wavelet transform architectures" over the convolution based one. Hence here we are using a lifting based 9/7 DWT.

Fig.3 illustrates the flipping structure of 9/7 DWT designed by *Haung*, explained in the article "Flipping structure: An efficient VLSI architecture for lifting based discrete wavelet transform (DWT)," for the purpose of lifting-based 1-D DWT. The flipping structure shares the identical computational complexity with the traditional lifting scheme. This flipping structure diminishes the critical path by flipping some computation units with the inverse of coefficients of the multiplier. Constants *C0.............C5* are given by

 $CI = 1/(\alpha\beta) = 0.7437502472$ 

 $C2 = 1/(\beta \gamma) = -0.6680671710$ 

 $C3 = 1/(\gamma \delta) = 0.6384438531$ 

 $C4 = \alpha \beta \delta / \zeta = 2.065244244$ 

 $C5 = \alpha \beta \gamma \delta \zeta = 2.421021152$ 

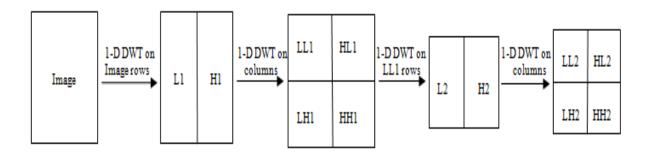


Fig 2: Two-Level Wavelet Decomposition.
The Dotted Portions Are the Final Wavelet Transformed Data.

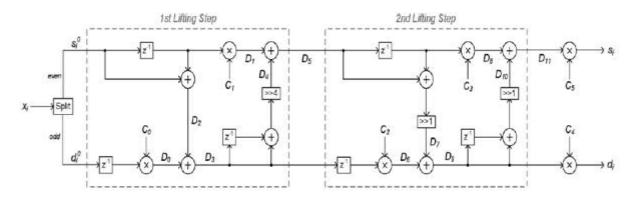


Fig 3: Bit Parallel Architecture

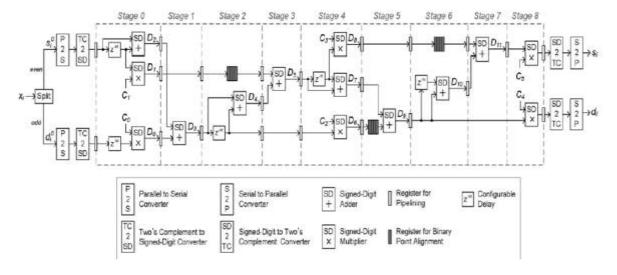


Fig 4: 1-D 9/7 DWT

#### 4.2 Quantization

Quantization engaged in image processing is a lossy compression technique. It can be accomplished by compressing a range of values to a single quantum value i.e., fixed value. One stream can be compressible by reducing the number of discrete symbols in that stream. For example, reduce the file size when you need to represent an image, in place of reducing the number of colours, we can go for it. Precise applications include DCT data quantization in JPEG and DWT data quantization in JPEG 2000.

The key element for the lossy 9/7 DWT when achieving compression performance is Quantization. As per the previous presentation, uniform dead-zone quantization has supported by the JPEG 2000 standard and also it supports Trellis coded quantization. Because of the simplicity and hardware efficiency, we are selecting Uniform dead-zone quantization.

This quantization procedure utilizes bins, which have identical in size except only for a "dead zone" quantizer centred at zero, has the size of a double of the others. For example, by using this quantization scheme, if n bits of precision for HL1, LH1 and HH1, then there should be the n+1 bits of precision for LL2, HL2, LH2 and HH2 as clarified from the presentation of "Precision for 2-D discrete wavelet transform processors".

#### V. PROPOSED ARCHITECTURE

It is desirable to minimize the bit-widths for all variables in the data paths, most important to reduction of size in tables, and operators, for instance adders and multipliers. Here we are using a bit width minimization scheme. It can able to minimize bit-widths while make sure that the results meet the one ulp error bound requirement. We have to split the problem of minimizing fixed-point bit-widths into two kinds of parts: range analysis followed by precision analysis. These two different parts are performed totally within MATLAB framework, making use of the finite precision hardware emulation models and an arithmetical approach is taken to undertake the range and precision minimization problems. Range analysis engages checking the dynamic range and totalling the bit-widths of the integer parts.

Overflows or underflows and excessive bits cause waste valuable hardware resources happen because of the insufficient bits for the range. Simulation-based approach can be used by the range analysis design is forwarding with a bulky set of random numbers. It ranges over not only the interval of possible values for the particular input, but also including the extreme values of that interval.

Coming to BP approach the primary goal is computing speed, the design challenge is determining the suitable number of integer and fractional bits to use in representing all the signals, which has utilized during the computation. For signal z, the number of integer bits are denoted by IBz, fractional bits are denoted by FBz and the total number of bits has denoted by Bz, where Bz = IBz + FBz.

#### **5.1 Integer Bit- Width Determination**

The IB can be determined by using an approach. The approach is based on computing the roots of the derivatives of each signal. Since the binary point needs to be associated for additions, the two addition operands have to share the same IB. As per this phenomena, for the 1-D DWT, which has shown in Fig. 3, the following signal pairs have to share the same IB, i.e., (D0,D2), (D1,D4), (D6,D7) and (D8,D10). Practically, this involves that the IB need to be set as the larger IB of the two. For example, IBD0 = IBD2 = (IBD0, IBD2). Moreover si, di are the final output data.

# 5.2 Fractional Bit-Width Optimization

The fractional bit-width optimization is carried out in two steps, one is a static step based on analytical models, for which to obtain the set of widths, and a dynamic step is based on simulation that further diminish the bit width using a PSNR delta threshold. The target precision metric (ulp) error criterion is one of the beat ways to specify the worst case (maximum absolute) error. The static step finds the set of bits that guarantee less than 2-ulp error at the final outputs of quantized DWT.

# 5.3 Static Optimization

The worst case (maximum absolute error) quantization errors for truncation and round-to-nearest are given by

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Truncation: Ez = max (0, 2-FBz - 2FBz') ...(1)
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Round-to-nearest: 
$$Ez = \{0, \text{ if } FBz \ge FBz' 2\text{-}FBz - 1, \text{ otherwise}\}\$$
 ... (2)

Where FBz' is the un-quantized full precision.

$$Esi = \max(D11) \times 2\text{-FBc5} - 1 + C5 \times ED11 + ED11 \times 2\text{-FBC5} - 1 + \max(0, 2\text{-FBsi } -2\text{-FBc5} - \text{FBD11})...(3)$$

$$Edi = \max(D9) \times 2\text{-FBC4} - 1 + \text{C4} \times ED9 + ED9 \times 2\text{-FBC4} - 1 + \max(0, 2\text{-FBdi} - 2\text{-FBc4} - \text{FBD9})...(4)$$

Here these error expressions consider as the worst case error bounds at each node and can be recursively obtained for any number of DWT levels. Internal data paths bit widths are create using the error expressions in conjunction with simulated annealing. Since the quantization scheme, which is of JPEG 2000 uses increasing precision with i=L.

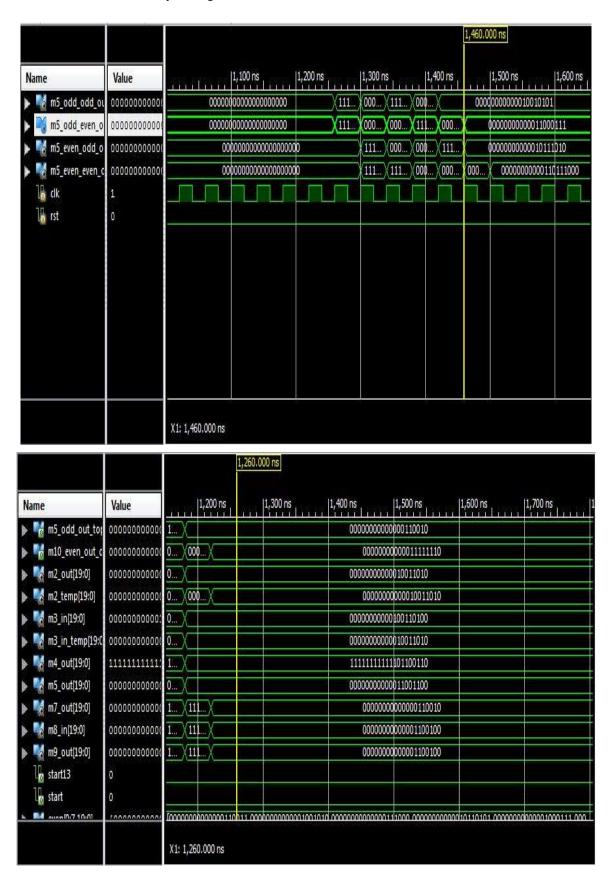
#### VI. DS DWT DESIGN

Fig. 4 illustrates the DS 1-D 9/7 DWT design. First, the incoming two's complement data is serialized. After that it has converted into SD representation. The serial SDs is after that passed into the DS DWT, which is divided into nine pipeline stages that run in parallel. After the previous stage, the DWT-transformed data is exchanged back into two's complement representation and afterward parallelized into words. This approach can have the ability to reduce the memory requirement. Because of the reason is two's complement occupies half the area of the equivalent SD representation. Both SD addition and SD multiplications creates one digit per cycle, where starting from the most significant digit used for the static step is the unit in the last place.

The radix-2 SD redundant number system is used in the DS representations. SD operations do not propagate carries because of redundancy. That's why they are able to run in most significant digit first (MSDF) mode. This MSDF property makes it attractive for the DS DWT approach since it allows for varying the number of iterations, which is to obtain different precision.

#### VII. IMPLEMENTATION RESULTS

We have designed this process in verilog HDL language. A main reason beyond using this language is easier than any other HDL language and because of some of its salient features like it can able to allow the descriptions of each module to completed mathematically in terms its terminals and external parameters functioned to the module.etc. The design has simulated using the standard simulation tool Xilinx. The results in terms of numbers and are analysed to get accurate results.



#	Value		1,260.000 ns							
Name		1,200 ns	1,300 ns  1,400 ns  1,500 ns  1,600 ns  1,700 ns							1,700 ns
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▶ 🕌 m5_odd_even_o	1111111111111	000000000000000000000000000000000000000	111	000	X000	X11L.	. \(000\)	0000	0000000011000111	
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• 🌃 m3_in[19:0]	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000	000	(000	X111	X	1111111	1111100111010	
• 🤻 m3_in_temp[19:0	000000000000000000000000000000000000000	000000000000000000000000000000000000000	000	000	(000	X11	X	1111111	1111110011101	
• 🌃 m4_out[19:0]	111111111111	000000000000000000000000000000000000000	(111	111	(111	X000	X	0000000	000001100011	
• 🌃 m5_out[19:0]	000000000000000000000000000000000000000	000000000000000000000000000000000000000	0000	000	(000	X111	X	0000000	000000110010	
• 🌃 m7_out[19:0]	000000000000000000000000000000000000000	000000000000000000000000000000000000000	0000	111	(000	X111	X000X	0000	0000000010010101	
• 🌃 m8_in[19:0]	111111111111	000000000000000000000000000000000000000	111	000	(000	X000	. X000X	0000	0000000100101010	
• 🌃 m9_out[19:0]	111111111111	000000000000000000000000000000000000000	111	000	(000	X000	X000X	0000	0000000100101010	
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#### VIII. CONCLUSION

In this paper, precision-aware approaches and associated hardware implementations for performing the DWT are presented. Both BP and DS design methodologies and also results have been showed. These methods can use of an optimal amount of hardware resources in the DWT computation. In addition, this framework enables quantization, which is usually performed after the DWT in algorithms such as JPEG 2000. We believe that the design methods and architectures such as those presented in this presentation plays significant role in the design of future energy- and precision-optimized DWT implementations.

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